



DC COUPLING HIGH VOLTAGE VIDEO AMPLIFIER

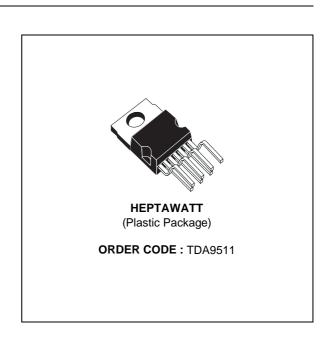
BANDWIDTH: 40MHz TYPICAL
RISE AND FALL TIME: 9ns TYPICAL

■ SUPPLY VOLTAGE: 110V ■ POWER DISSIPATION: 3.0W

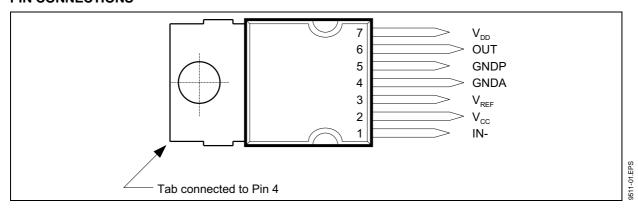
■ ESD PROTECTED



The TDA9511 is a video amplifier designed with a high voltage Bipolar/CMOS/DMOS technology (BCD). It drives in DC coupling mode one cathode of a monitor and is protected against flashovers. It is available in Heptawatt package.



PIN CONNECTIONS

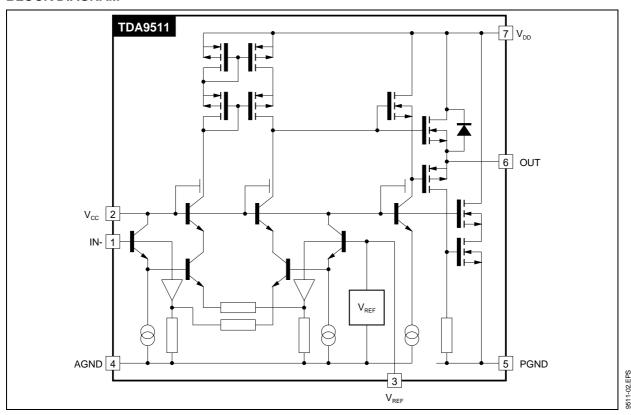


PIN CONFIGURATION

Pin N	Symbol	Function		
1	IN-	Input of the amplifier		
2	V _{CC}	Low Voltage Power Supply (12V Typ.)		
3	V_{REF}	Internal Voltage Reference (3.3V)	1	
4	GNDA	Analog Ground	1	
5	GNDP	Power Ground	1	
6	OUT	Output driving the cathode	1.TBL	
7	V_{DD}	High Voltage Power Supply (110V Max.)	9511-01.TBI	

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BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{DD}	Supply High Voltage (Pin 7)	120	V
V _{CC}	Supply Low Voltage (Pin 2)	20	V
VESD	ESD Susceptibility Human Body Model, 100pF Discharge through 1.5k Ω EIAJ Norm, 200pF Discharge through 0Ω	2 300	kV V
I _{OD} I _{OG}	Output Current to V _{DD} (Pin 6) Output Current to Ground (Pin 6) (see Note 1)	protected 80	mA
lj	Input Current (Pin 1)	50	mA
Tj	Junction Temperature	150	°C
T _{oper}	Operating Ambient Temperature	0, +70	°C
T _{stg}	Storage Temperature	-20, +150	°C

Note 1 : Pulsed current $t \leq 50 \mu s$

THERMAL DATA

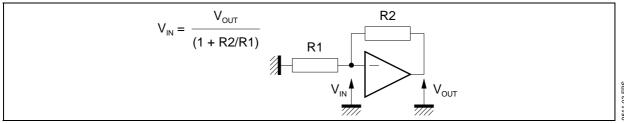
Symbol	Parameter	Value	Unit
R _{th (j-c)}	Junction-Case Thermal Resistance Max.	3	°C/W
R _{th (j-a)}	Junction-Ambient Thermal Resistance Typ.	70	°C/W

ELECTRICAL CHARACTERISTICS (V_{CC} = 12V, V_{DD} = 110V, T_{amb} = 25°C, unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V_{DD}	High Supply Voltage (Pin 7)		20		110	V
Vcc	Low Supply Voltage (Pin 2)		10	12	15	V
I _{DD}	DC Current of High Voltage Supply (without feedback current)	V _{OUT} = 60V		9		mA
Icc	Low Voltage Supply Internal DC Current			15		mA
V_{REF}	Internal Reference (Pin 3)			3.2		V
V_{IN}	Input Voltage	V _{OUT} = 60V		3.25		V
$dV_{\text{IN}}/dV_{\text{CC}}$	Drift of Input Voltage versus V _{CC}	Measured on Pin 1		0.12		%
dV _{IN} /dT	Drift of Input Voltage versus Temperature			0.5		mV/°C
V _{SATH}	High Output Saturation Voltage (Pin 6)	I _O = -60mA		V _{DD} - 8.5		V
V _{SATL}	Low Output Saturation Voltage (Pin 6)	I _O = 60mA		12		V
ELin	Linearity Error	17V < V _{OUT} < V _{DD} - 15V			5	%
OS	Overshoot			5		%
BW	Bandwidth at -3dB	$\begin{array}{l} \text{Measured on CRT cathodes.} \\ \text{C}_{\text{LOAD}} = 10 \text{pF, Rprotect} = 220 \Omega, \\ \text{V}_{\text{OUT}} = 60 \text{V, } \Delta \text{V}_{\text{OUT}} = 20 \text{V}_{\text{PP}}, \\ \text{Feedback gain} = 20 \end{array}$		40		MHz
t _R , t _F	Rise and Fall Time	$\label{eq:measured_between 10\% \& 90\%} \begin{tabular}{ll} Measured between 10\% \& 90\% \\ of output pulse, \\ C_{LOAD} = 10pF, Rprotect = 220\Omega, \\ V_{OUT} = 60V, \Delta V_{OUT} = 40V_{PP} \end{tabular}$		9		ns
Go	Open Loop Gain	V _{OUT} = 60V		60		dB
	Open Loop Gain Temperature Coefficient			0.03		dB/°C
I _{IB}	Input Bias Current (Pin 1)	V _{OUT} = 60V		20	30	μΑ
	Input Bias Temperature Coefficient	_		90		nA/°C
R _{IN}	Input Resistance	See Note 2		200		kΩ

Note 2: Characterized and not tested.

Figure 1: Measurement of Input Voltage



TYPICAL APPLICATION

The TDA9511 consists of:

- A differential amplifier with active load,
- A DMOS output buffer,
- Abandgap voltage reference (Pin 3 for filtering only).

PC board lay-out

The best performances are obtained with a carefully designed HF PC-Board, especially for the output and input capacitors.

The feedback resistor R_F must have a low parasitic capacitor ($C_F < 0.3pF$).

This parasitic capacitor C_F must be compensated by a capacitor R3 (roughly 20 · C_F) connected in parallel with the input resistor R1.

The full bandwidth of the device is only obtained with well matched compensation otherwise the application will have either an integrator response with a low bandwidth or a differentiator response with too much ringing.

A diode D_P (see Figure 2) has to be connected for flashover protection.

Power dissipation

The power dissipation consists of a static part and a dynamic part. The static dissipation varies with the output voltage and the feedback resistor. The dynamic power dissipation increases with the pixel frequency.

For a signal frequency of 40MHz and $40V_{PP}$ output signal, the typical power dissipation is about 3.0W, for $V_{DD} = 110V$.

In first approximation, the dynamic dissipation is:

$$P_D = V_{DD} * C_{LOAD} * \Delta V_{OUT} * f$$

and the total dissipation is:

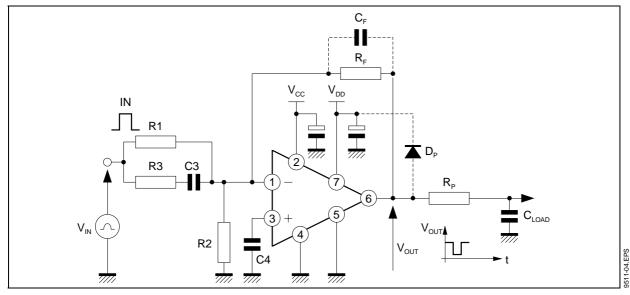
$$\mathsf{P} = \mathsf{V}_{\mathsf{DD}} * \mathsf{C}_{\mathsf{LOAD}} * \Delta \mathsf{V}_{\mathsf{OUT}} * \mathsf{f} + \mathsf{V}_{\mathsf{DD}} * \mathsf{I}_{\mathsf{DD}}$$

+
$$V_{CC}$$
 * I_{CC} - $(V_{DD}$ - $\overline{V_{OUT}})$ $\frac{\overline{V_{OUT}}}{R_{FEEDBACK}}$

with f = pixel frequency

P = 110V x 10pF x 40V x 40MHz + 110V x 7mA +12 x 20mA - 60^2 V/20kΩ = 2.95W

Figure 2: Typical Evaluation Schematic



Recommended values :

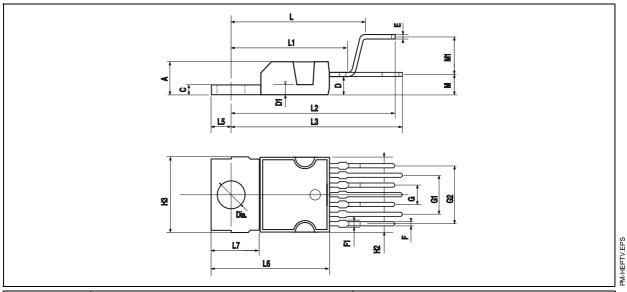
 $R1 = 1k\Omega$, $R2 = 1.8k\Omega$, $R_F = 20k\Omega$, $R_P = 200\Omega$,

C4 > 10nF, C3 = 10 to 12pF for $C_F # 0.5pF$.

R3 # 150Ω .



PACKAGE MECHANICAL DATA: 7 PINS - PLASTIC HEPTAWATT



Dimensions	Millimeters			Inches			
	Min.	Тур.	Max.	Min.	Тур.	Max.	
Α			4.8			0.189	
С			1.37			0.054	
D	2.4		2.8	0.094		0.110	
D1	1.2		1.35	0.047		0.053	
E	0.35		0.55	0.014		0.022	
F	0.6		08	0.024		0.031	
F1			0.9			0.035	
G	2.41	2.54	2.67	0.095	0.100	0.105	
G1	4.91	5.08	5.21	0.193	0.200	0.205	
G2	7.49	7.62	7.8	0.295	0.300	0.307	
H2			10.4			0.409	
H3	10.05		10.4	0.396		0.409	
L		16.97			0.668		
L1		14.92			0.587		
L2		21.54			0.848		
L3		22.62			0.891		
L5	2.6		3	0.102		0.118	
L6	15.1		15.8	0.594		0.622	
L7	6		6.6	0.236		0.260	
М		2.8			0.110		
M1		5.08			0.200		
Dia.	3.65		3.85	0.144		0.152	

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