APPLICATION MANUAL

LDO REGULATOR WITH ON/OFF SWITCH TK112xxCM/U



Features

- Very low Dropout Voltage. (Vdrop=105mV at 100mA)
- Very good stability (CL=0.1 μ F is stable for any type capacitor with 2.5V \leq Vout)
- High Precision output Voltage ($\pm 1.5\%$ or ± 50 mV)
- Good ripple rejection ratio (80dB at 1kHz)
- Wide operating voltage range $(1.8V \sim 14.5V)$
- Peak output current is 480mA.(10% down point)
- Built-in Short circuit protection
- Built-in Thermal Shutdown
- Suitable for Very Low Noise Applications
- Built-in on/off Control (0.1µA Max Standby current) High On
- Very Small Surface Mount Packages SOT23L / SOT89 package
- Built-in reverse bias over current protection

Description

The TK112xxC is an integrated circuit with a silicon monolithic bipolar structure. The regulator is of the low saturation voltage output type with very little quiescent current $(65\mu A)$.

The PNP power transistor is built-in. The I/O voltage difference is 0.17V (typical) when a current of 200mA is supplied to the system. Because of the low voltage drop, the voltage source can be effectively used; this makes it very suitable for battery powered equipment.

The on/off function is built into the IC. The current during standby mode becomes very small (pA level).

The output voltage is available from 1.5 to 10.0V in 0.1V steps. The output voltage is trimmed with high accuracy. This allows the optimum voltage to be selected for the equipment.

The over current sensor circuit and the reverse-bias protection circuit are built-in.

It is a very rugged design because the ESD protection is high. Therefore, the TK112xxC can be used with confidence.

When mounted on the PCB, the power dissipation rating becomes about 600mW/ 900mW, even though the packages are very small.

The TK112xxC features very high stability in both DC and AC.

The capacitor on the output side provides stable operation with $0.1\mu F$ with $2.5V \le Vout$. A capacitor of any type can be used; however, the larger this capacitor is, the better the overall characteristics are.

ORDERING INFORMATION

T K 1 1 2 [C [
Voltage Code (Refer to the following table) Version C	Environment Code -GH: Lead Free and Halogen Free (SOT23L-6) -G: Lead Free (SOT89-5) Tape / Reel Code L: Left Type (SOT23L-6)
Package Code	B : Back Type (SOT89-5)
M : SOT23L-6	Rank Code
U : SOT89-5	C : C Rank
	I : I Rank

Voltage Code

V OUT	V CODE	V OUT	V CODE	V OUT	V CODE	V OUT	V CODE
1.5	15	2.5	25	3.5	35	4.5	45
1.6	16	2.6	26	3.6	36	4.6	46
1.7	17	2.7	27	3.7	37	4.7	47
1.8	18	2.8	28	3.8	38	4.8	48
1.9	19	2.9	29	3.9	39	4.9	49
2.0	20	3.0	30	4.0	40	5.0	50
2.1	21	3.1	31	4.1	41		
2.2	22	3.2	32	4.2	42		
2.3	23	3.3	33	4.3	43		
2.4	24	3.4	34	4.4	44		



Absolute Maximum Ratings

Ta=25°C

Parameter	Symbol	Rating	Units	Conditions
Absolute Maximum Ratings				
Supply Voltage	Vcc_{MAX}	- 0.4 ~ 16	V	
Reverse Bias	Vrov	- 0.4 ~ 6	V	Vout ≤ 2.0V
Reverse bias	Vrev _{MAX}	- 0.4 ∼ 12	V	2.1V ≤ Vout
Np pin Voltage	Vnp _{MAX}	- 0.4 ∼ 5	V	
Control pin Voltage	Vcont _{MAX}	- 0.4 ~ 16	V	
Storage Temperature Range	T_{stg}	- 55 ∼ 150	°C	
Power Dissipation	P_D	SOT23L-6: 600 SOT89-5: 900	mW	Internal Limited Tj=150°C *
Operating Condition				
Operating Temperature Range	T_{OP}	-4 0 ~ 85	°C	
Operating Weltage Pange	V	2.1 ~ 14.5	V	$T_{OP} = -40 \sim 85^{\circ}C$
Operating Voltage Range	V_{OP}	1.8 ~ 14.5	V	$T_{OP} = -30 \sim 80^{\circ}C$
Short Circuit Current	Ishort	500	mA	

^{*} P_D must be decreased at rate of 4.8mW/°C(SOT23L-6), 7.2mW/°C(SOT89-5) for operation above 25°C. The maximum ratings are the absolute limitation values with the possibility of the IC being damaged. If the operation exceeds any of these standards, quality cannot be guaranteed.

Electrical Characteristics

(1) C rank

The operation between $-40 \sim 85^{\circ}$ C is guaranteed by design. The parameter with limit value will be guaranteed with test when manufacturing or SQC (Statistical Quality Control) technique.

Vin=Vout_{TYP}+1V,Vcont=1.8V,Ta=25°C

Parameter	Symbol		Value		Units	Candi	itions
rarameter	Symbol	MIN	TYP	MAX	Units	Conditions	
Output Voltage	Vout	Refe	er to TABI	LE 1	V	Iout = 5mA	
Line Regulation	LinReg	-	0.0	6.0	mV	$\Delta Vin = 5V$	
Load Regulation	LoaReg	Refe	er to TABI	LE 1	mV	$Iout = 5mA \sim 10$	0mA
		Refe	er to TABI	LE 1	mV	$Iout = 5mA \sim 20$	0mA
		Refe	er to TABI	LE 1	mV	$Iout = 5mA \sim 30$	0mA
Dropout Voltage *1	Vdrop	-	105	170	mV	Iout = 100mA	
		-	170	270	mV	Iout = 200mA	
		-	235	370	mV	Iout = $270mA$ (2.	$1 \text{V} \le \text{Vout} \le 2.3 \text{V}$
		-	235	370	mV	Iout = $300mA$ (2.	4V ≤ Vout)
Maximum Output Current *2	Iout _{MAX}	380	480	-	mA	When (Vout _{TYP} ×	0.9)
Supply Current	Iq	-	65	90	μΑ	Iout = 0mA	
Standby Current	Istandby	-	0.0	0.1	μΑ	Vcont = $0V$	
Quiescent Current	Ignd	-	1.8	3.0	mA	Iout = 100mA	
Control Terminal							
Control Current	Icont	-	5.0	10	μΑ	Vcont = 1.8V	
Control Voltage	Vcont	1.8	-	-	V	Vout ON state	T = 40 959C
		-	-	0.35	V	Vout OFF state	T_{OP} =-40~85°C
		1.6	-	-	V	Vout ON state	T = 20, 80°C
		-	-	0.6	V	Vout OFF state	T_{OP} =-30~80°C

^{*1:} For Vout ≤ 2.0 V, no regulations.

^{*2:} The maximum output current is limited by power dissipation.



TABLE 1. Output Voltage, Load Regulation

	0.	ıtput Volta	100	Load Regulation						
Dant Number) Oi	ութաւ ν οπε	ige	Iout = 100mA						
Part Number	MIN	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
	V	V	V	mV	mV	mV	mV	mV	mV	
TK11213C	1.250	1.300	1.350	11	24	21	49	34	77	
TK11214C	1.350	1.400	1.450	11	24	22	49	34	78	
TK11215C	1.450	1.500	1.550	11	24	22	50	35	79	
TK11216C	1.550	1.600	1.650	11	24	22	50	35	80	
TK11217C	1.650	1.700	1.750	11	25	22	51	36	82	
TK11218C	1.750	1.800	1.850	11	25	23	51	36	83	
TK11219C	1.850	1.900	1.950	11	25	23	52	37	84	
TK11220C	1.950	2.000	2.050	11	25	23	53	37	85	
TK11221C	2.050	2.100	2.150	11	26	23	53	38	86	
TK11222C	2.150	2.200	2.250	12	26	24	54	38	88	
TK11223C	2.250	2.300	2.350	12	26	24	54	39	89	
TK11224C	2.350	2.400	2.450	12	26	24	55	39	90	
TK11225C	2.450	2.500	2.550	12	27	24	55	40	91	
TK11226C	2.550	2.600	2.650	12	27	25	56	40	92	
TK11227C	2.650	2.700	2.750	12	27	25	56	41	93	
TK11228C	2.750	2.800	2.850	12	27	25	57	41	95	
TK11229C	2.850	2.900	2.950	12	27	25	58	42	96	
TK11230C	2.950	3.000	3.050	12	28	26	58	42	97	
TK11231C	3.050	3.100	3.150	12	28	26	59	43	98	
TK11232C	3.150	3.200	3.250	12	28	26	59	44	99	
TK11233C	3.250	3.300	3.350	13	28	26	60	44	101	
TK11234C	3.349	3.400	3.451	13	29	27	60	45	102	
TK11235C	3.447	3.500	3.553	13	29	27	61	45	103	
TK11236C	3.546	3.600	3.654	13	29	27	62	46	104	
TK11237C	3.644	3.700	3.756	13	29	27	62	46	105	
TK11238C	3.743	3.800	3.857	13	29	28	63	47	107	
TK11239C	3.841	3.900	3.959	13	30	28	63	47	108	
TK11240C	3.940	4.000	4.060	13	30	28	64	48	109	
TK11241C	4.038	4.100	4.162	13	30	28	64	48	110	
TK11242C	4.137	4.200	4.263	13	30	29	65	49	111	
TK11243C	4.235	4.300	4.365	14	31	29	66	49	112	
TK11244C	4.334	4.400	4.466	14	31	29	66	50	114	
TK11245C	4.432	4.500	4.568	14	31	29	67	50	115	
TK11246C	4.531	4.600	4.669	14	31	30	67	51	116	
TK11247C	4.629	4.700	4.771	14	31	30	68	51	117	
TK11248C	4.728	4.800	4.872	14	32	30	68	52	118	
TK11249C	4.826	4.900	4.974	14	32	30	69	52	120	
TK11250C	4.925	5.000	5.075	14	32	31	70	53	121	

TABLE 1. Output Voltage , Load Regulation (continue)

	Output Voltage			Output Voltage Load Regulation							
Part Number				Iout =	100mA	Iout =	200mA	nA Iout = 300mA			
r art Number	MIN	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX		
	V	V	\mathbf{V}	mV	mV	mV	mV	mV	mV		
TK11251C	5.023	5.100	5.177	14	32	31	70	53	121		
TK11253C	5.220	5.300	5.380	15	33	31	71	54	124		
TK11254C	5.319	5.400	5.481	15	33	32	72	55	125		
TK11255C	5.417	5.500	5.583	15	33	32	72	55	127		
TK11260C	5.910	6.000	6.090	15	34	33	75	58	133		
TK11280C	7.880	8.000	8.120	17	39	38	87	68	156		



(2) I rank

The operation between $-40 \sim 85$ °C is guaranteed with normal test. The parameter with limit value will be guaranteed with test when manufacturing or SQC(Statistical Quality Control) technique.

 $Vin=Vout_{TYP}+1V,Vcont=1.8V,Ta=-40 \sim 85^{\circ}C$

Parameter	Symbol	Value I		Units	Conditions	
rarameter	Symbol	MIN	TYP	MAX	Units	Conditions
Output Voltage	Vout	Refe	er to TABl	LE 1	V	Iout = 5mA
Line Regulation	LinReg		0.0	8.0	mV	$\Delta Vin = 5V$
Load Regulation	LoaReg	Refe	er to TABl	LE 1	mV	$Iout = 5mA \sim 100mA$
		Refe	er to TABl	LE 1	mV	$Iout = 5mA \sim 200mA$
		Refe	er to TABl	LE 1	mV	$Iout = 5mA \sim 300mA$
Dropout Voltage *1	Vdrop		105	200	mV	$Iout = 100mA (2.2V \le Vout)$
			170	320	mV	$Iout = 200mA (2.2V \le Vout)$
			235	440	mV	$Iout = 300mA (2.4V \le Vout)$
Maximum Output Current *2	Iout _{MAX}	340	480		mA	When (Vout _{TYP} ×0.9)
Supply Current	Iq		65	100	μΑ	Iout = 0mA
Standby Current	Istandby		0.0	0.5	μΑ	Vcont = 0V
Quiescent Current	Ignd		1.8	3.6	mA	Iout = 100mA
Control Terminal						
Control Current	Icont		5.0	12	μΑ	Vcont = 1.8V
Control Voltage	Vcont	1.8			V	Vout ON state
				0.35	V	Vout OFF state

^{*1}: For Vout ≤ 2.1V , no regulations.

^{*2:} The maximum output current is limited by power dissipation.

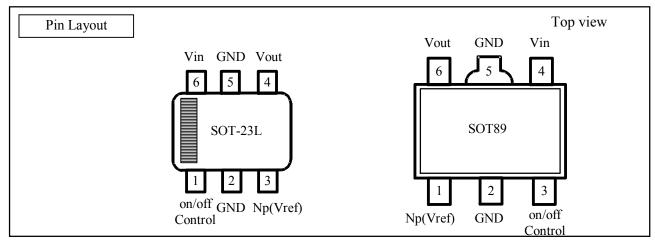


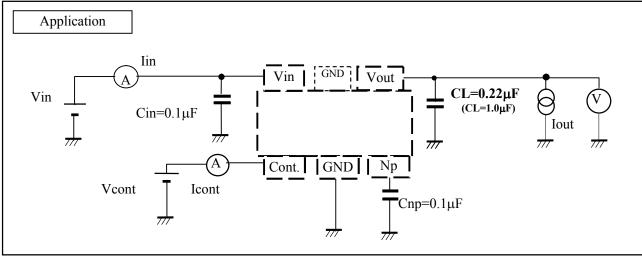
TABLE 1. Output Voltage, Load Regulation

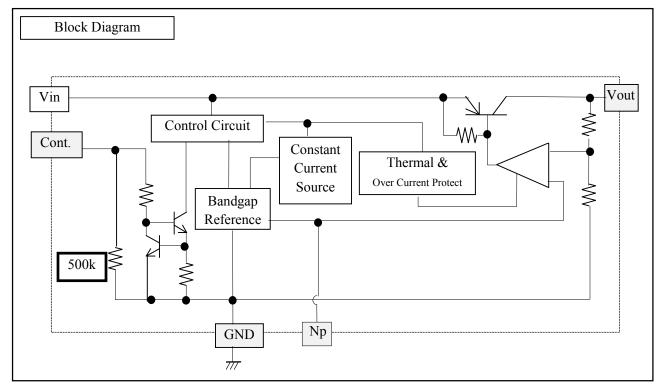
	Output Voltage			Load Regulation						
Part Number	Oi	ութաւ νοπ	ige	Iout =	100mA	200mA	nA Iout = 300mA			
I alt Number	MIN	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX	
	V	V	\mathbf{V}	mV	mV	mV	mV	mV	mV	
TK11213C	1.220	1.300	1.380	11	29	21	60	34	95	
TK11214C	1.320	1.400	1.480	11	29	22	61	34	96	
TK11215C	1.420	1.500	1.580	11	29	22	61	35	97	
TK11216C	1.520	1.600	1.680	11	29	22	62	35	98	
TK11217C	1.620	1.700	1.780	11	30	22	63	36	100	
TK11218C	1.720	1.800	1.880	11	30	23	63	36	118	
TK11219C	1.820	1.900	1.980	11	30	23	64	37	120	
TK11220C	1.920	2.000	2.080	11	30	23	65	37	122	
TK11221C	2.020	2.100	2.180	11	31	23	65	38	124	
TK11222C	2.120	2.200	2.280	12	31	24	66	38	126	
TK11223C	2.220	2.300	2.380	12	31	24	67	39	127	
TK11224C	2.320	2.400	2.480	12	31	24	68	39	129	
TK11225C	2.420	2.500	2.580	12	31	24	68	40	131	
TK11226C	2.520	2.600	2.680	12	32	25	69	40	133	
TK11227C	2.620	2.700	2.780	12	32	25	70	41	135	
TK11228C	2.720	2.800	2.880	12	32	25	70	41	137	
TK11229C	2.820	2.900	2.980	12	32	25	71	42	139	
TK11230C	2.920	3.000	3.080	12	33	26	72	42	141	
TK11231C	3.020	3.100	3.180	12	33	26	73	43	143	
TK11232C	3.120	3.200	3.280	12	33	26	73	44	145	
TK11233C	3.217	3.300	3.383	13	33	26	74	44	147	
TK11234C	3.315	3.400	3.485	13	33	27	75	45	149	
TK11235C	3.412	3.500	3.588	13	34	27	75	45	151	
TK11236C	3.510	3.600	3.690	13	34	27	76	46	153	
TK11237C	3.607	3.700	3.793	13	34	27	77	46	155	
TK11238C	3.705	3.800	3.895	13	34	28	77	47	157	
TK11239C	3.802	3.900	3.998	13	34	28	78	47	159	
TK11240C	3.900	4.000	4.100	13	35	28	79	48	161	
TK11241C	3.997	4.100	4.203	13	35	28	80	48	162	
TK11242C	4.095	4.200	4.305	13	35	29	80	49	164	
TK11243C	4.192	4.300	4.408	14	35	29	81	49	166	
TK11244C	4.290	4.400	4.510	14	36	29	82	50	168	
TK11245C	4.387	4.500	4.613	14	36	29	82	50	170	
TK11246C	4.485	4.600	4.715	14	36	30	83	51	172	
TK11247C	4.582	4.700	4.818	14	36	30	84	51	174	
TK11248C	4.680	4.800	4.920	14	36	30	84	52	176	
TK11249C	4.777	4.900	5.023	14	37	30	85	52	178	
TK11250C	4.875	5.000	5.125	14	37	31	86	53	180	

TABLE 1. Output Voltage , Load Regulation (continue)

	Output Voltage		Load Regulation						
Part Number			Iout =	100mA	Iout =	200mA	Iout =	300mA	
Tart Number	MIN	TYP	MAX	TYP	MAX	TYP	MAX	TYP	MAX
	V	V	V	mV	mV	mV	mV	mV	mV
TK11255C	5.362	5.500	5.638	15	38	32	89	55	190
TK11257C	5.557	5.700	5.843	15	38	32	91	56	194
TK11260C	5.850	6.000	6.150	15	39	33	93	58	199
TK11280C	7.800	8.000	8.200	17	43	38	107	68	238





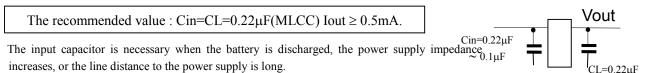




Input /Output Capacitors

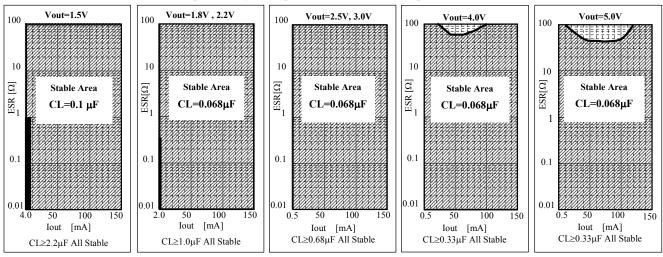
Linear regulators require input and output capacitors in order to maintain the regulator's loop stability. If a $0.1\mu F$ capacitor is connected to the output side, the IC provides stable operation at any voltage in the practical current region. However, increase the CL capacitance when using the IC in the low current region and low voltage. Otherwise, the IC oscillates.

The equivalent series resistance (ESR) of the output capacitor must be in the stable operation area. However, it is recommended to use as large a value of capacitance as is practical. The output noise and the ripple noise decrease as the capacitance value increases. ESR values vary widely between ceramic and tantalum capacitors. However, tantalum capacitors are assumed to provide more ESR damping resistance, which provides greater circuit stability. This implies that a higher level of circuit stability can be obtained by using tantalum capacitors when compared to ceramic capacitors with similar values.



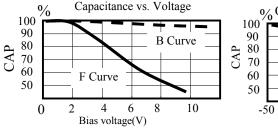
This capacitor might be necessary on each individual IC even if two or more regulator ICs are used. It is not possible to where this indiscriminately. Please confirm the stability while mounted. The IC provides stable operation with an output side capacitor of $0.1\mu F$ (Vout $\geq 2.5V$). If it is $0.1\mu F$ or more over the full range of temperature, either a ceramic capacitor or tantalum capacitor can be used without considering ESR. It is not possible to say indiscriminately. Please confirm stability while mounted.

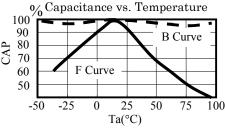
Output voltage, Output current vs. Stable Operation Area



The above graphs show stable operation with a ceramic capacitor of 0.1uF (excluding the low current region). If the capacitance is not increased in the low voltage, low current area, stable operation may not be achieved. Please select the best output capacitor according to the voltage and current used. The stability of the regulator improves if a big output side capacitor is used (the stable operation area extends.) Please use as large a capacitance as is practical. Although operation above 150 mA has not been described, stability is equal to or better than operation at 150 mA.

For evaluation Kyocera :CM05B104K10AB , CM05B224K10AB , CM105B104K16A , CM105B224K16A , CM21B225K10A Murata :GRM36B104K10 , GRM42B104K10 , GRM39B104K25 , GRM39B224K10 , GRM39B105K6.3

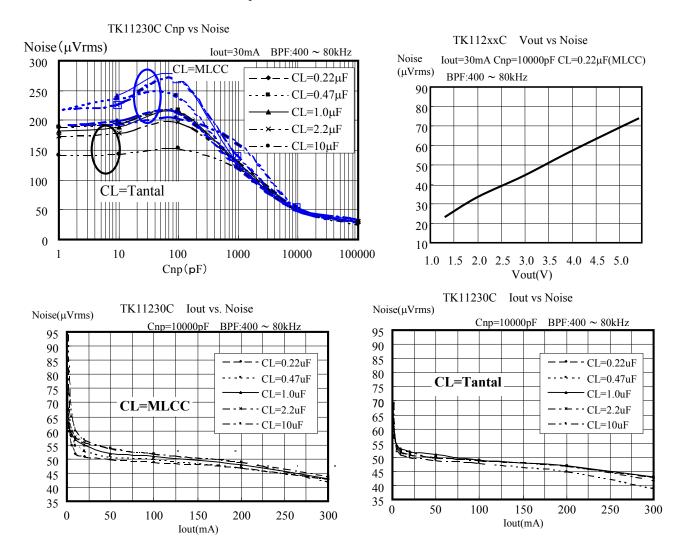




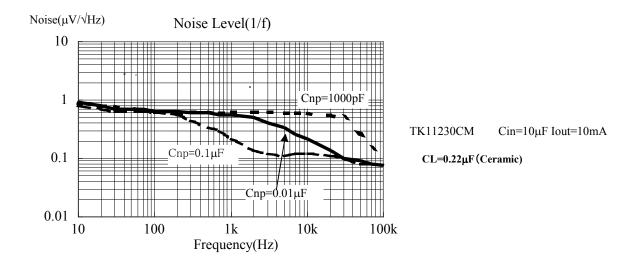
Generally, a ceramic capacitor has both a temperature characteristic and a voltage characteristic. Please consider both characteristics when selecting the part. The B curves are the recommend characteristics.

Output noise

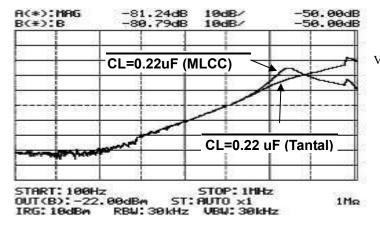
TK11230C Cnp vs. Noise Iout=30mA BPF=400Hz ~ 80kHz



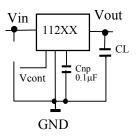
Increase Cnp to decrease the noise. The recommended Cnp capacitance is $6800 pF(682) \sim 0.22 \mu F(224)$. The amount of noise increases with the higher output voltages.

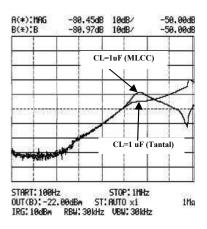


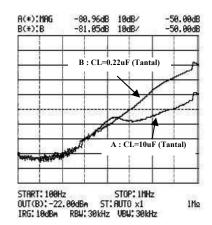
Ripple rejection

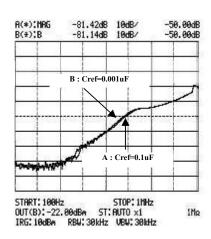


 $\label{eq:Vin=5.0V} Vin=5.0V \quad Vout=3.0V \quad Iout=10mA \\ VR=500mVp-p \quad f=100 \sim 1MHz \quad Cnp=0.1uF$

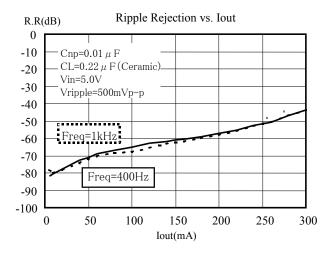


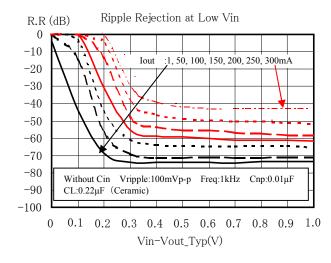






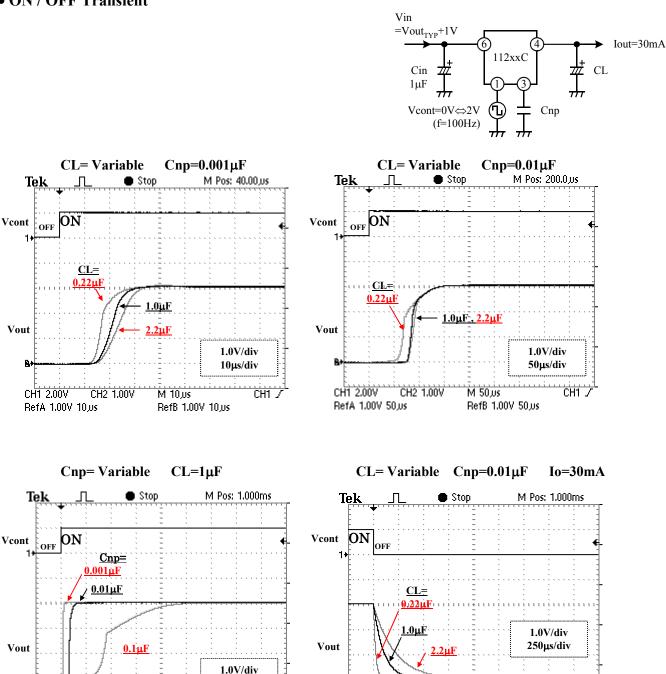
The ripple rejection characteristic depends on the characteristic and the capacitance value of the capacitor connected to the output side. The RR characteristic of 50KHz or more varies greatly with the capacitor on the output side and PCB pattern. If necessary, please confirm stability while operating.





TK112xxC Transient

• ON / OFF Transient



The rise time of the regulator depends on CL and Cnp; the fall time depends on CL.

250µs/div

M 250,us

RefB 1.00V 250,us

CH2 1.00V

CH1 2.00V

RefA 1.00V 250,us

CH1 2,00V CH2 1,00V

RefA 1.00V 250,us

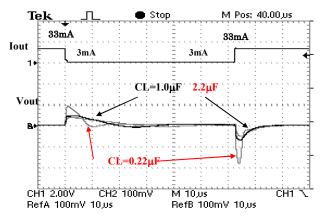
M 250 us

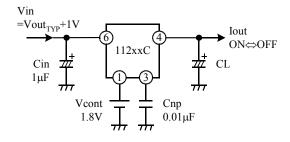
RefB 1.00V 250.us

CH1

• LOAD Transient

CL= Variable Cnp=0.01µF





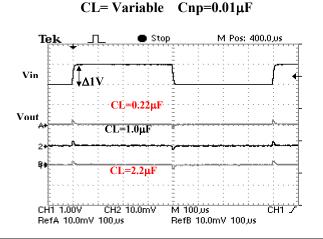
When the capacitor on the load side is increased, the load change becomes smaller.

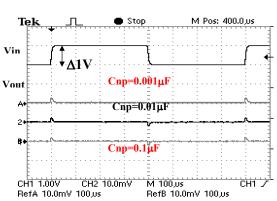
Magnification Iout= $0 \Leftrightarrow 30 \text{mA}$, $3 \Leftrightarrow 33 \text{mA}$ M Pos: 40.00,us M Pos: 4.000ms Iout=0 ⇔30mA Iout=0⇔30mA . Vout Vout Iout=3 ⇔ 33mA Iout=3⇔33mA CH2 200m M 10.us CH1 2.00% RefA 200mV 10,us CH1 2.00V CH2 200mV M 1ms CH1 RefA 200mV 1ms

The no load voltage change can be greatly improved by delivering a little load current to ground (see right curve above).

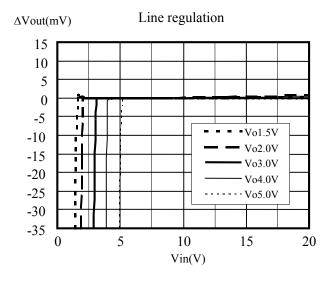
Increase the load side capacitor when the load change is fast or when there is a large current change. In addition, at no load, the voltage change can be reduced by delivering a little load current to ground.

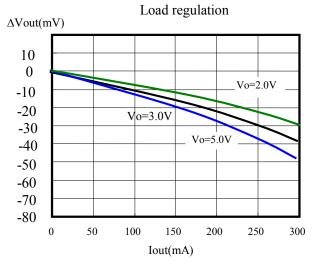
• Line Transient

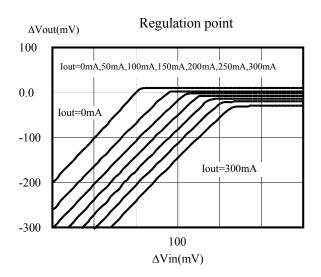


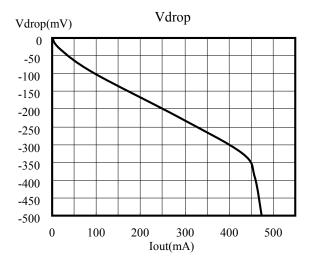


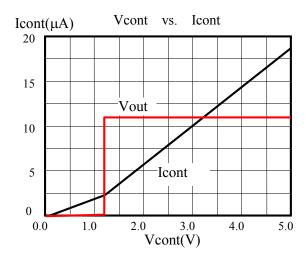
Cnp= Variable CL=1µF

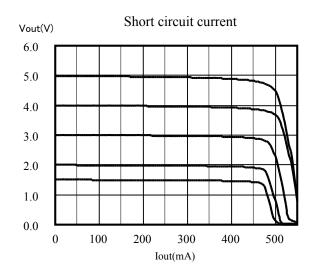


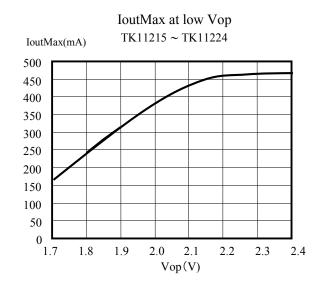


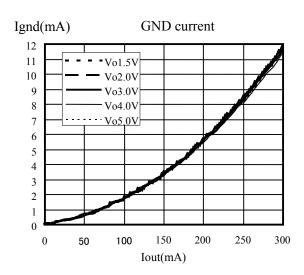


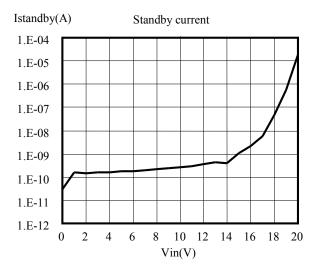


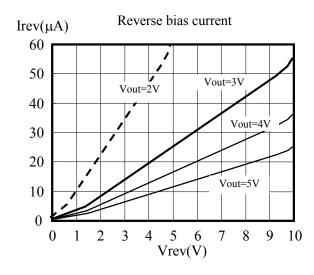




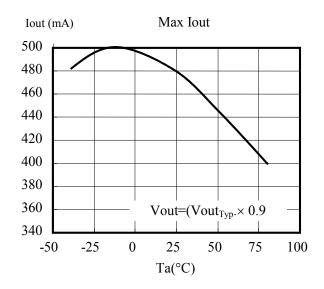




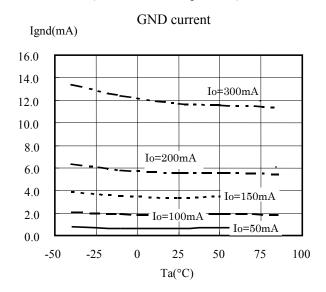


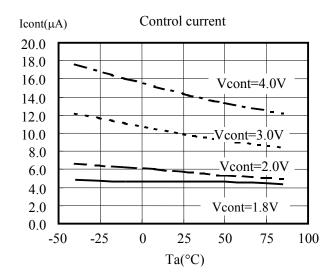


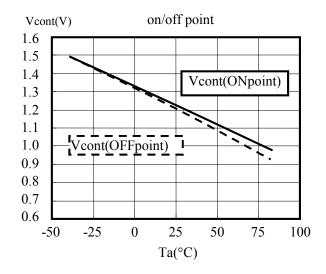
Temperature Characteristics

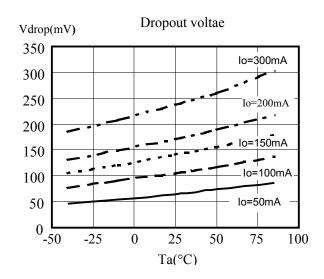


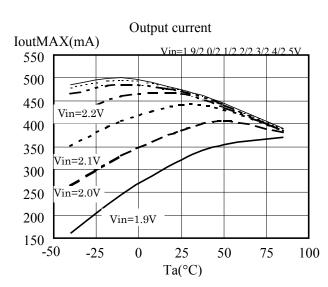
(Ta: Ambient temperature)



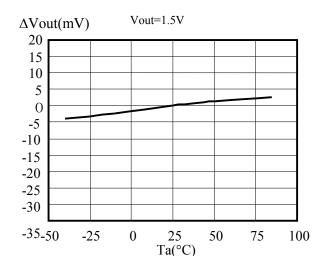


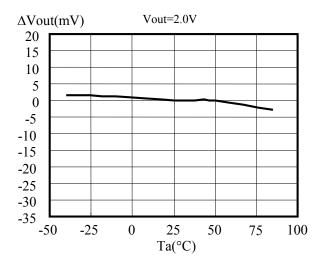


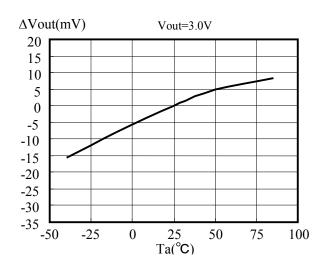


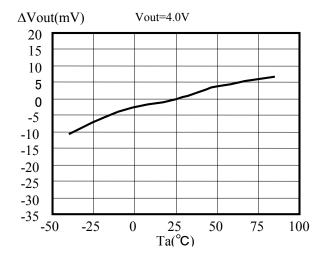


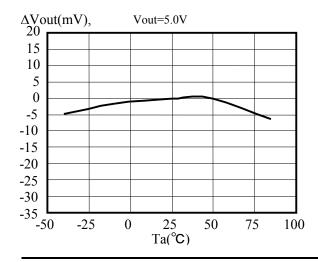
Output voltage vs. Temperature characteristics











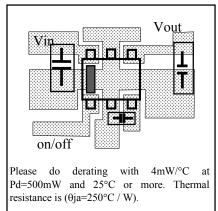


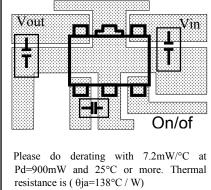
Layout PCB Material : Glass epoxy t=0.8mm

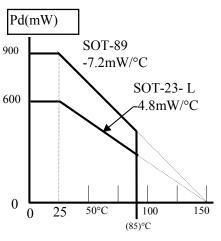
SOT-23L

SOT-89

Derating Curve







The package loss is limited at the temperature that the internal temperature sensor works (about 150°C). Therefore, the package loss is assumed to be an internal limitation. There is no heat radiation characteristic of the package unit assumed because of the small size. Heat is carried away by the device being installed on the PCB. This value changes by the material and the copper pattern etc. of the PCB. The losses are approximately 600mW (SOT-23L): 900mW(SOT-89). Enduring these losses becomes possible in a lot of applications operating at 25°C.

Determining the thermal resistance when mounted on a PCB.

The operating chip junction temperature is shown by

 $Ti=\theta ia \times Pd + Ta$. Ti of the IC is set to about 150°C.

Pd is a value when the overtemperature sensor is made to work.

Ta (Ta=25°C)

$$150 = \theta ja \times pd + 25$$

 $\theta ja \times Pd = 125$
 $\theta ja = (125/pd) (°C/mW)$

Pd is easily obtained.

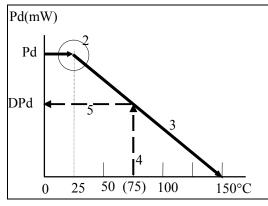
Mount the IC on the PCB. Pd becomes Vin × Iin when the output side of the IC is short-circuited.

The input current decreases gradually by the temperature rise of the chip.

Please use the value when the current is steady (thermal equilibrium is reached).

In many cases, heat radiation is good, and Pd becomes 600mW/900 mW or more.

Pd is obtained by the normal temperature in degrees. The current that can be used at the highest operating temperature is obtained from the graph of the figure below.



Procedure (Do when PCB mounted).

- 1. Pd is obtained ($Vin \times Iin$ when the output side is short-circuited).
- 2. Pd is plotted on the horizontal line to 25°C.
- 3. Pd is connected with the point of 150°C by the straight line (bold face line).
- 4. A line is extended vertically above the point of use temperature in the design. For instance, 75°C is assumed (broken line).
- 5. Extend the intersection of the derating curve (fat solid line) and (broken line) to the left and read the Pd value.
- 6. DPd ÷ (Vinmax − Vout)=Iout (at 75°C)

The maximum current that can be used at the highest operating temperature is:

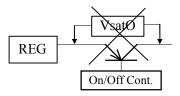
Iout \cong DPd \div (Vinmax – Vout).



Application hint

On/Off Control

It is recommended to turn the regulator Off when the circuit following the regulator is non-operating. A design with little electric power loss can be implemented. We recommend the use of the on/off control of the regulator without using a high side switch to provide an output from the regulator. A highly accurate output voltage with low voltage drop is obtained.



Because the control current is small, it is possible to control it directly by CMOS logic.

The PULLDOWN resistance (500K Ω) is built into the control terminal.

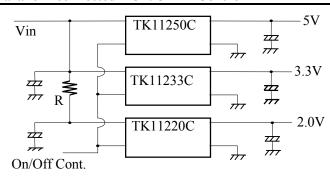
The noise and the ripple rejection characteristics depend on the capacitance on the Vref terminal.

The ripple rejection characteristic of the low frequency region improves by increasing the capacitance of Cnp.

A standard value is $Cnp=0.068\mu F$. Increase Cnp in a design with important output noise and ripple rejection requirements. The IC will not be damaged if the capacitor value is increased.

The on/off switching speed changes depending on the Np terminal capacitance. The switching speed slows when the capacitance is large.

Parallel connected ON/OFF Control



The figure at the left illustrates multiple regulators being controlled by a single On/Off control signal. There is a possibility of overheating because the power loss of the low voltage side IC (TK11220C) is large. The series resistor (R) is put in the input line of the low output voltage regulator in order to prevent over-dissipation. The voltage dropped across the resistor reduces the large input-to-output voltage across the regulator, reducing the power dissipation in the device. When the thermal sensor works, a decrease of the output voltage, oscillation, etc. may be observed.



Definition of Terms

The output voltage tables are specified with a test voltage of Vin=Output Voltage (Typ.) + 1V.

Output Voltage (Vout)

The output voltage is specified with (Vin = Output Voltage (Typ.) + 1V) and output current (Iout=5mA).

Maximum Output Current (Iout Max)

The output current is measured when the output voltage decreases to $(Vout_{Typ.} \times 0.9)$. The input voltage is (Output Voltage (Typ.) + 1V). The maximum output current is measured in a short time so that it is not influenced by the temperature of the chip. The output current decreases with low voltage operation.

Please refer to the "Low input voltage-output current" graph for 2.1V or less.

Dropout Voltage (Vdrop)

The dropout voltage is the difference between the input voltage and the output voltage at which point the regulator starts to fall out of regulation. Below this value, the output voltage will fall as the input voltage is reduced. It is dependent upon the load current (Iout) and the junction temperature (Tj). The input voltage is gradually decreased below the test voltage. It is the voltage difference between the input and the output when the output voltage decreases by 100mV.

Line Regulation (Lin Reg)

Line regulation is the ability of the regulator to maintain a constant output voltage as the input voltage changes. The line regulation is specified as the input voltage is changed from (Output Voltage (Typ.) + 1V) to (Output Voltage (Typ.) + 6V). This measurement is not influenced by the temperature of the IC and is measured in a short time.

Load Regulation (Load Reg)

Load regulation is the ability of the regulator to maintain a constant output voltage as the load current changes. The input voltage is set to (Output Voltage (Typ.) + 1V). The output voltage change is measured as the load current changes from to 5 to 100mA and from 5 to 200mA. This measurement is not influenced by the temperature of the IC and is measured in a short time.

Quiescent Current (Iq)

The quiescent current is the current which flows through the ground terminal under no load conditions (Io=0mA).

Ripple Rejection (RR)

Ripple rejection is the ability of the regulator to attenuate the ripple content of the input voltage at the output. It is specified with the input voltage = (Vout + 1.5V) , Iout=10mA, CL=1.0 μ F and Cnp=0.01 μ F An Alternating Current source of (f=1kHz and 200mV_{RMS}) is superimposed to the power-supply voltage. Ripple rejection is the ratio of the ripple content of the output vs. the input and is expressed in dB. It is typically about 80dB at 1KHz. The ripple rejection improves when the value of the capacitor at the noise bypass terminal in the circuit is large. However, the on/off response worsens.

Standby Current.(Istandby)

Standby current is the current which flows into the regulator when the control voltage is made 0 volts. It is measured with an input voltage of 8V.



PROTECTION CIRCUITS

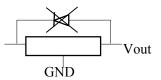
Thermal Sensor

The thermal sensor protects the device if the junction temperature exceeds the safe value (Tj=150 °C). This temperature rise can be caused by extreme heat, excessive power dissipation caused by large output voltage drops, or excessive output current. The regulator will shut off when the temperature exceeds the safe value. As the junction temperature decreases, the regulator will begin to operate again. Under sustained fault conditions, the regulator output will oscillate as the device turns off then resets. Please improve heat radiation or lower the input electric power. When heat radiation is poor, the forecast package loss is not obtained.

* In the case that the power, Vin × Ishort(Short Circuit Current), becomes more than twice of the maximum rating of its power dissipation in a moment, there is a possibility that the IC is destroyed before internal thermal protection works.

Reverse Bias Current

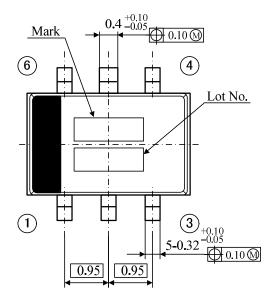
The reverse bias protection prevents excessive current from flowing through the IC even if the input voltage becomes 0 with voltage impressed on the output side (input short-circuited to GND). The maximum reverse bias voltage is 6V.

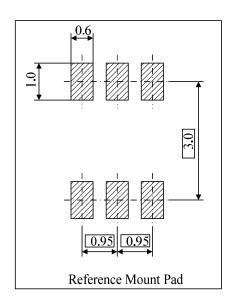


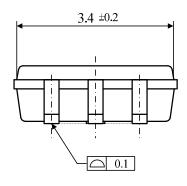
• ESD MM 200pF 0Ω 200V Min HBM 100pF 1.5k Ω 2000V Min

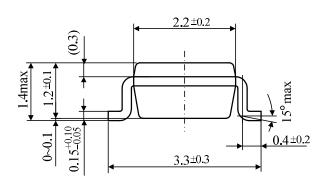
Outline; PCB; Stamps

SOT23L-6





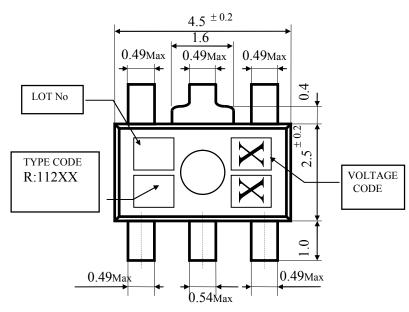


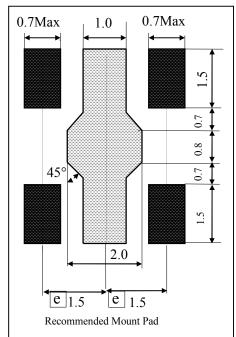


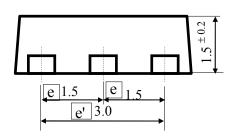
Unit: mm

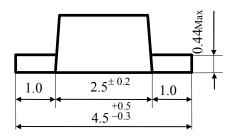
General tolerance : ± 0.2

SOT89-5









Unit: mm

General tolerance : ± 0.2

Asahi KASEI TK112xxCM/U

1. NOTES

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 - Medical devices for measuring blood pressure, pulse, etc., treatment units such as coronary pacemakers and heat treatment units, and devices such as artificial organs and artificial limb systems which augment physiological functions.
 - Electrical instruments, equipment or systems used in disaster or crime prevention.
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ASAHI KASEI TOKO POWER DEVICES CORPORATION

13-45, Senzui 3-chome, Asaka-shi, Saitama-ken

351-0024, Japan

TEL: +81-48-460-1870 (Marketing Department)

FAX: +81-48-460-1600

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